

General description

The MX16171D100/ MX16171S100 high-side OR-ing FET controller works with an external MOSFET and acts as an ideal diode rectifier when connected in series with the power supply. This OR-ing controller enables MOSFETs to replace diode rectifiers in power distribution networks, reducing power loss and voltage drop.

The MX16171D100/ MX16171S100 controller provides charge pump gate drive for an external N-channel MOSFET and fast response comparator to turn off the FET when current flows in reverse. The MX16171D100/ MX16171S100 can be connected to power supplies from 1V to 100V (a separate VS supply is needed when IN is 1V to 4V) and can withstand transient voltages up to 110V.

Features

- ◆ Wide operating input voltage range V_{IN} : 5V to 100V
- ◆ 110V transient voltage
- ◆ Charge pump gate driver for external N-channel MOSFET
- ◆ 50ns fast response to current reversal
- ◆ 2A peak gate off current
- ◆ Ultra-small V_{DS} turn-off voltage reduces turn-off time
- ◆ DFN2*3-8L and SOP8

Applications

Active OR-ing of redundant (N+1) power supplies

General information

Ordering information

Part Number	Description
MX16171D100	DFN2*3-8L
MX16171S100	SOP8
MPQ	3000pcs

Package dissipation rating

Package	R θ JA (°C/W)
DFN2*3-8L	65
SOP8	150

Absolute maximum ratings

Parameter	Value
IN, OUT Pins to GND	-0.3 to 100V
GATE Pin to GND	-0.3 to 110V
VS Pin to Ground	-0.3 to 100V
OFF Pin to Ground	-0.3 to 7V
Junction temperature	150°C
Storage temperature, Tstg	-55 to 150°C
Leading temperature (soldering, 10secs)	260°C
ESD Susceptibility HBM	±2000V

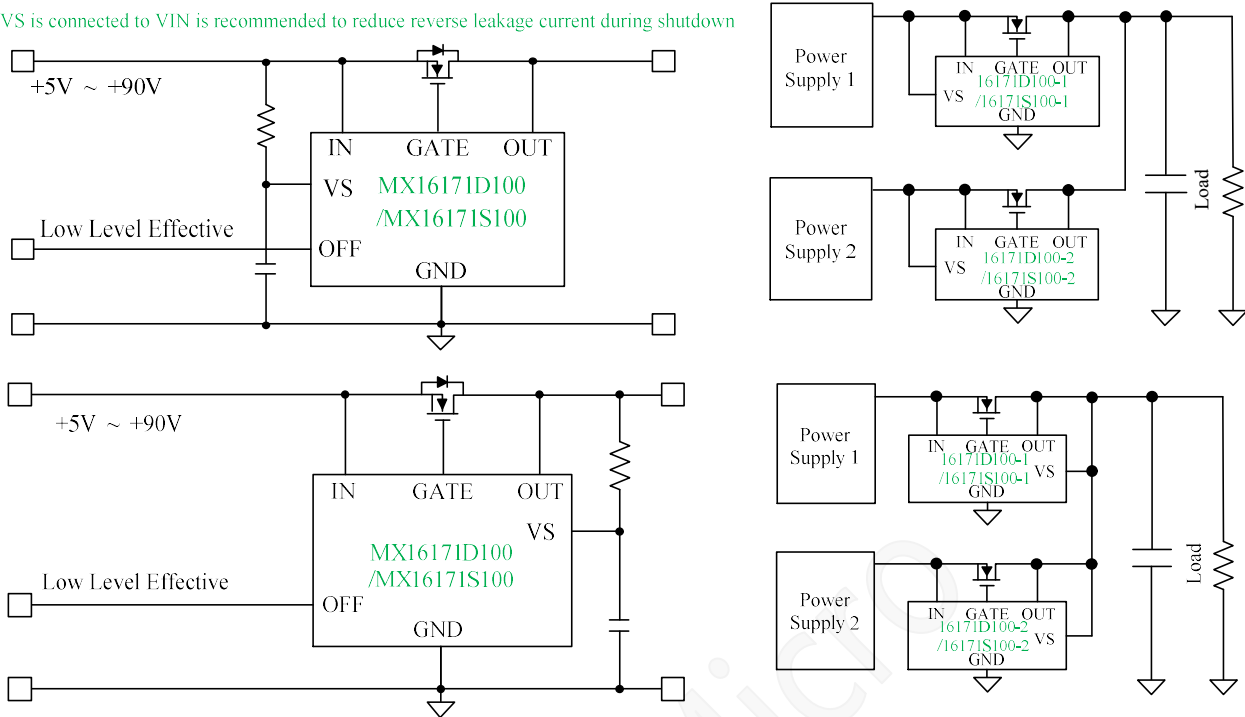
Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

Recommended operating condition

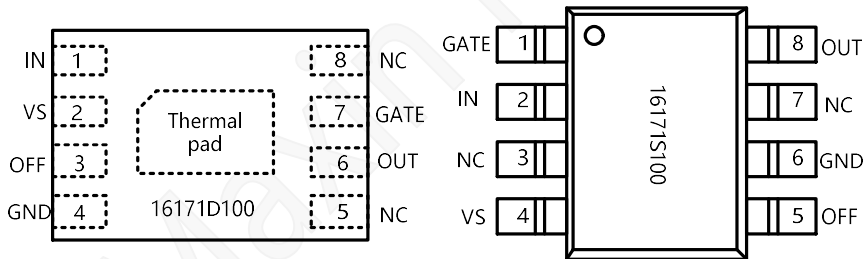
Symbol	Range
IN, OUT Pins ($V_S \geq 4.5V$ for $IN < 4V$)	1-90V
VS Pin	5-90V
OFF Pin	0-5.5V
Operating temperature	-40~125°C
Moisture sensitive level	MSL3

Typical application

VS is connected to VIN is recommended to reduce reverse leakage current during shutdown

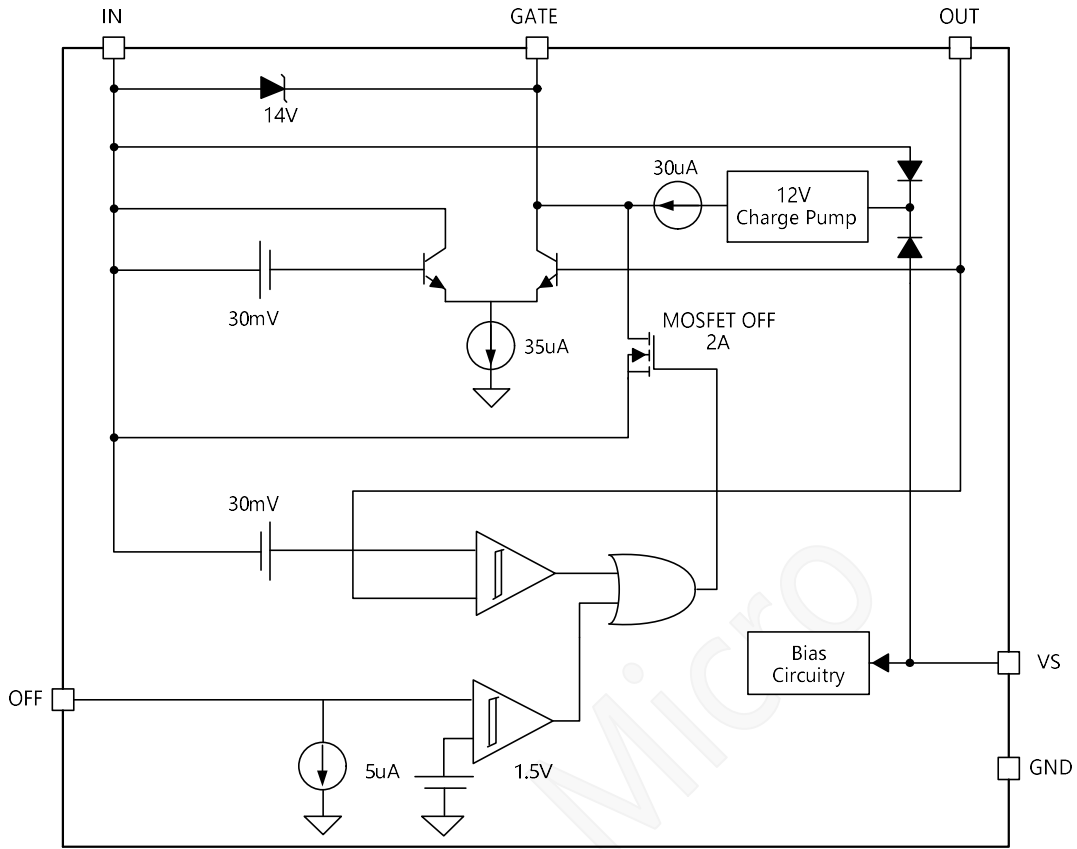


Terminal assignments

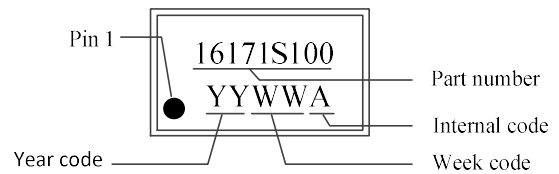
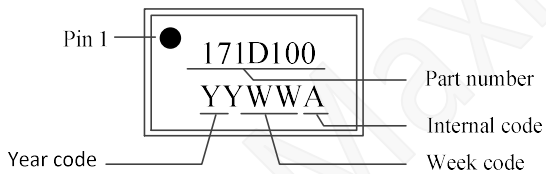


PIN NO.		PIN name	Description
DFN	SOP		
1	2	IN	Voltage sense connection to the external MOSFET Source pin.
2	4	VS	The main supply pin for all internal biasing and an auxiliary supply for the internal gate drive charge pump. Typically connected to either V_{OUT} or V_{IN} ; a separate supply can also be used. VS is connected to V_{IN} is recommended to reduce leakage current during reverse shutdown mode.
3	5	OFF	A logic high state at the OFF pin will pull the GATE pin low and turn off the external MOSFET. Note that when the MOSFET is off, current will still conduct through the FET's body diode. This pin should may be left open or connected to GND if unused.
4	6	GND	Ground return for the controller
5、8	3、7	NC	
6	8	OUT	Voltage sense connection to the external MOSFET Drain pin.
7	1	GATE	Connect to the Gate of the external MOSFET. Controls the MOSFET to emulate a low forward-voltage diode.
Thermal pad		Thermal pad	This pad can be connected to GND.

Block diagram

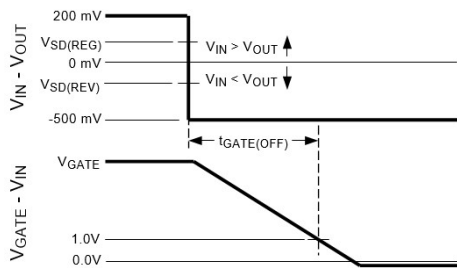


Marking information

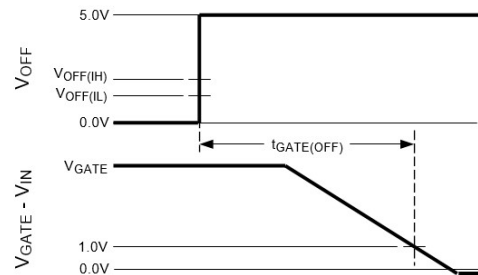


Electrical characteristics

($V_{IN} = 12-90V$, $V_{VS} = V_{IN}$, $V_{OUT} = V_{IN}$, $V_{OFF} = 0V$, $C_{GATE} = 47nF$, $T_A = 25^\circ C$, unless otherwise noted)



Gate OFF Timing for Forward to Reverse Transition



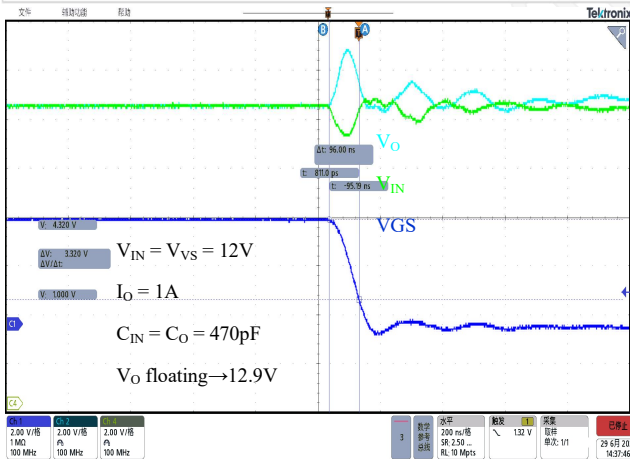
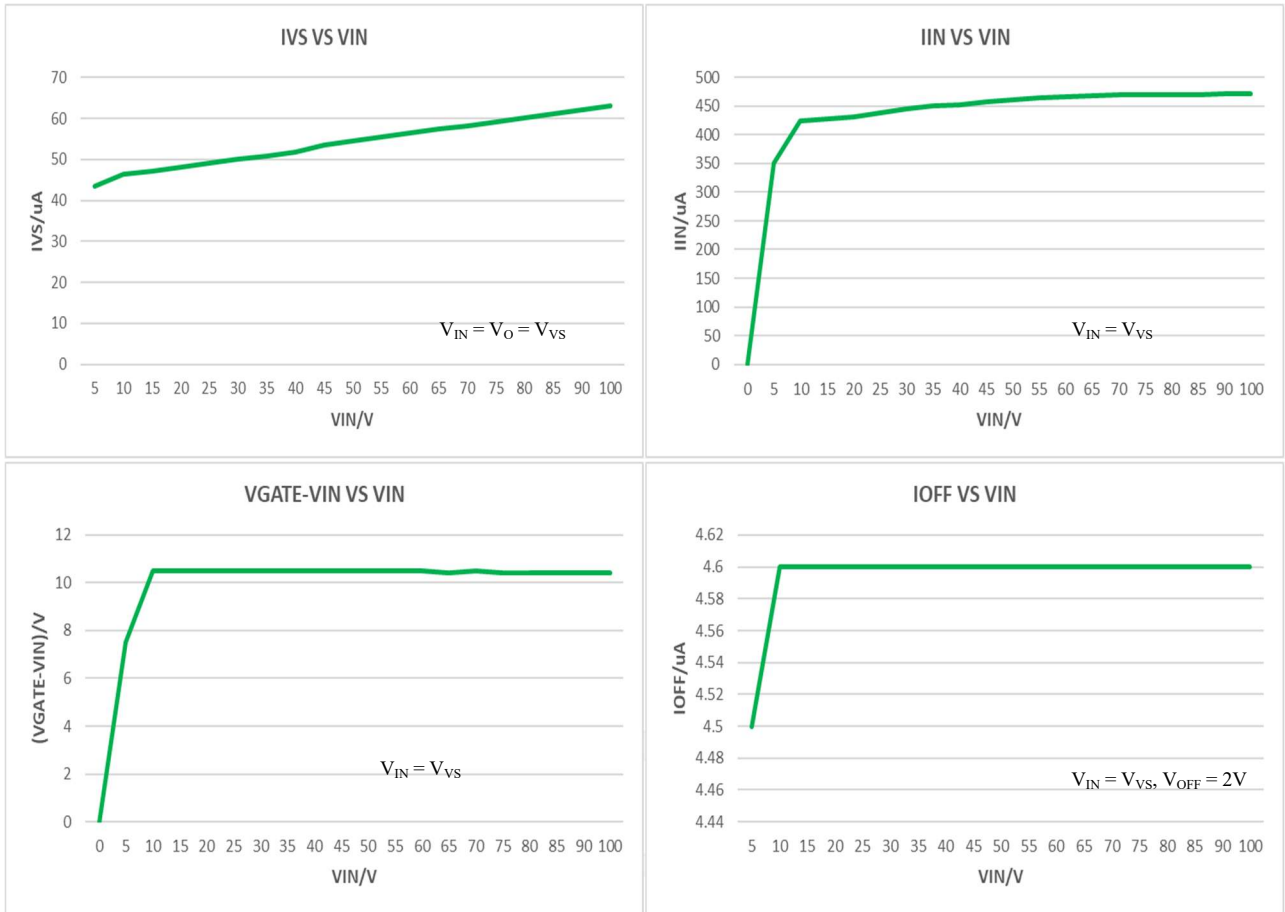
Gate OFF Timing for OFF Pin Low to High Transition

100V Input High-Side OR-ing FET Controller

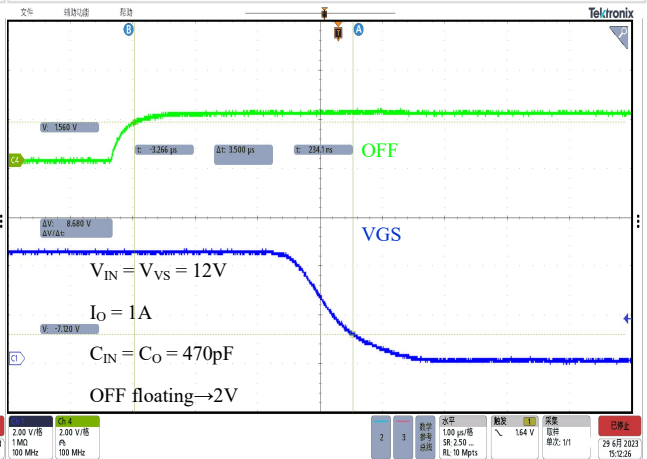
Symbol	Parameter	Test condition	Min	Typ.	Max	Unit
VS PIN						
V _{VS}	Operating Supply Voltage Range		5		90	V
I _{VS}	Operating Supply Current	V _{VS} = 5V, V _{IN} = 5V, V _{OUT} = V _{IN} - 100mV		50	70	uA
		V _{VS} = 12V, V _{IN} = 12V, V _{OUT} = V _{IN} - 100mV		55	70	
		V _{VS} = 40V, V _{IN} = 40V, V _{OUT} = V _{IN} - 100mV		60	100	
		V _{VS} = 90V, V _{IN} = 90V, V _{OUT} = V _{IN} - 100mV		65	110	
IN PIN						
V _{IN}	Operating Input Voltage Range		5		95	V
I _{IN}	IN Pin current	V _{IN} = 10V, V _{VS} = V _{IN} , V _{OUT} = V _{IN} - 100mV, GATE = Open	200		500	uA
		V _{IN} = 12V to 90V, V _{VS} = V _{IN} , V _{OUT} = V _{IN} - 100mV, GATE = Open	250		550	
OUT PIN						
V _{OUT}	Operating Output Voltage Range		5		95	V
I _{OUT}	OUT Pin Current	V _{IN} = 5V to 90V, V _{VS} = V _{IN} , V _{OUT} = V _{IN} - 100mV		3.2	8	uA
GATE PIN						
I _{GATE(ON)}	Gate Pin Source Current	V _{IN} = 5V, V _{VS} = V _{IN} , V _{GATE} = V _{IN} , V _{OUT} = V _{IN} - 175mV	-37		-25	uA
		V _{IN} = 12V to 100V, V _{VS} = V _{IN} , V _{GATE} = V _{IN} , V _{OUT} = V _{IN} - 175mV	-39		-25	
V _{GS}	V _{GATE} - V _{IN} in Forward Operation	V _{IN} = 5V, V _{VS} = V _{IN} , V _{OUT} = V _{IN} - 175mV	5	7	8	V
		V _{IN} = 8V to 100V, V _{VS} = V _{IN} , V _{OUT} = V _{IN} - 175mV	8	11	14	
t _{GATE(REV)}	Gate Capacitance Discharge Time at Forward to Reverse Transition	C _{GATE} = 0		300		ns
		C _{GATE} = 10nF		400		
		C _{GATE} = 47nF		500		
t _{GATE(OFF)}	Gate Capacitance Discharge Time at OFF pin Low to High Transition	C _{GATE} = 47nF		4.2		us
I _{GATE(OFF)}	Gate Pin Sink Current	V _{GATE} = V _{IN} + 3V, V _{OUT} > V _{IN} + 100mV, t ≤ 10ms		2		A
V _{SD(REV)}	Reverse V _{SD} Threshold V _{IN} < V _{OUT}	V _{IN} - V _{OUT}	-40	-15	-10	mV
ΔV _{SD(REV)}	Reverse V _{SD} Hysteresis			10		mV
V _{SD(REG)}	Regulated Forward V _{SD} Threshold V _{IN} > V _{OUT}	V _{IN} = 5V, V _{VS} = V _{IN} , V _{IN} - V _{OUT}	1	15	30	mV
		V _{IN} = 12V, V _{VS} = V _{IN} , V _{IN} - V _{OUT}	5	25	60	
OFF PIN						
V _{OFF(IH)}	OFF Input High Threshold Voltage	V _{OUT} = V _{IN} - 500mV, V _{OFF} Rising		1.55	1.8	V
V _{OFF(IL)}	OFF Input Low Threshold Voltage	V _{OUT} = V _{IN} - 500mV, V _{OFF} Falling	1.3	1.45		
ΔV _{OFF}	OFF Threshold Voltage Hysteresis	V _{OFF(IH)} - V _{OFF(IL)}		150		mV
I _{OFF}	OFF Pin Internal Pulldown	V _{OFF} = 5V	3	5	8	uA

Characteristic plots

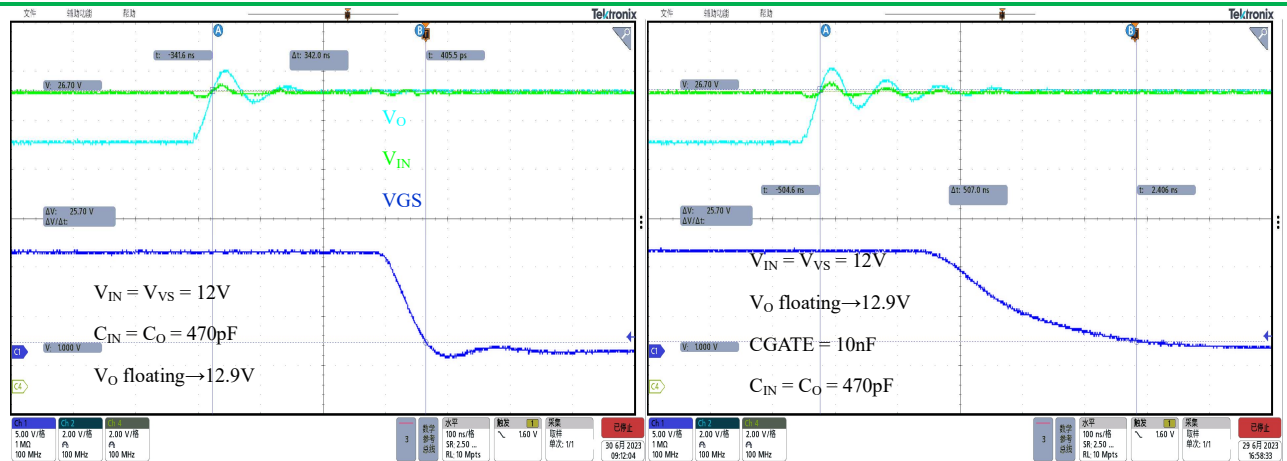
($V_{IN} = V_{VS} = V_{OUT}$, $V_{OFF} = 0V$, $C_{IN} = C_O = 470pF$, $T_A = 25^\circ C$, unless otherwise noted)



Reverse C_{GATE} discharge time

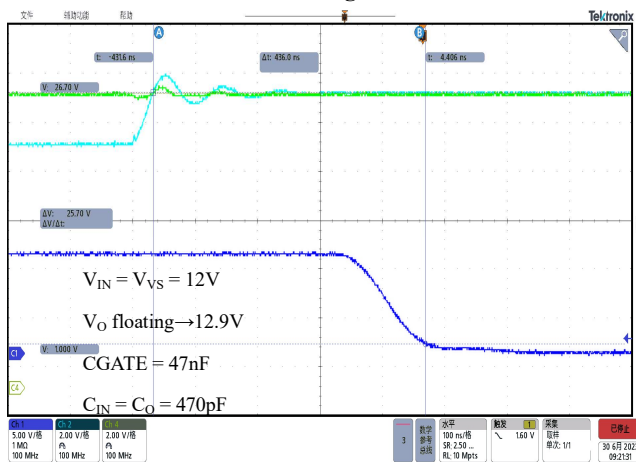


C_{GATE} discharge time VS OFF pin

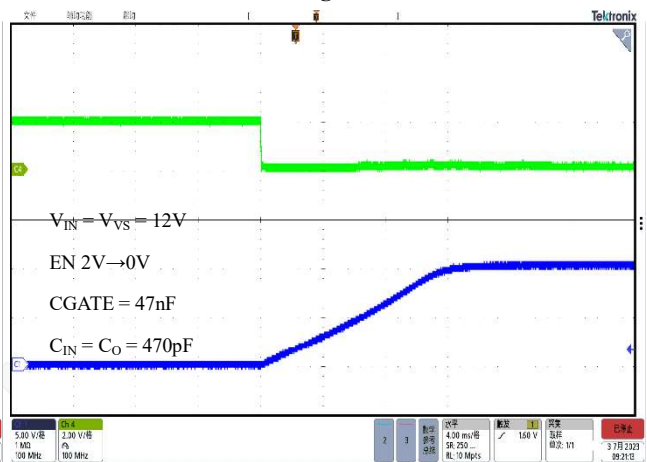


Reverse GATE discharge time

Reverse GATE discharge time, $C_{GATE} = 10nF$



Reverse GATE discharge time, $C_{GATE} = 47nF$



C_{GATE} charge time VS OFF pin

Maxin

Operation description

IN, GATE, and OUT Pins

When power is initially applied, the load current will flow from source to drain through the body diode of the MOSFET. Once the voltage across the body diode exceeds $V_{SD(REG)}$ then the MX16171D100/ MX16171S100 begins charging the MOSFET gate through a 30 μ A (typical) charge pump current source. In forward operation, the gate of the MOSFET is charged until it reaches the clamping voltage of the 14V GATE to IN pin Zener diode internal to the MX16171D100/ MX16171S100.

The MX16171D100/ MX16171S100 is designed to regulate the MOSFET gate-to-source voltage. If the MOSFET current decreases to the point that the voltage across the MOSFET falls below the $V_{SD(REG)}$ voltage regulation point of 30mV (typical), the GATE pin voltage will be decreased until the voltage across the MOSFET is regulated at 20mV. If the source-to-drain voltage is greater than the $V_{SD(REG)}$ voltage, the gate-to-source voltage will increase and eventually reach the 14V GATE to IN pin Zener clamp level.

If the MOSFET current reverses, possibly due to failure of the input supply, such that the voltage across the MX16171D100/ MX16171S100 IN and OUT pins is more negative than the $V_{SD(REV)}$ voltage of -15mV (typical), the MX16171D100/ MX16171S100 will quickly discharge the MOSFET gate through a strong GATE to IN pin discharge transistor. If the input supply fails abruptly, as would occur if the supply was shorted directly to ground, a reverse current will temporarily flow through the MOSFET until the gate can be fully discharged. This reverse current is sourced from the load capacitance and from the parallel connected supplies. The MX16171D100/ MX16171S100 responds to a voltage reversal condition typically within 50ns. The actual time required to turn off the MOSFET will depend on the charge held by the gate capacitance of the MOSFET being used. A MOSFET with 47nF of effective gate capacitance can be turned off in typically 260ns. This fast turnoff time minimizes voltage disturbances at the output, as well as the current transients from the redundant supplies.

VS Pin

The VS pin of MX16171D100/ MX16171S100 is the main supply pin for all internal biasing and an auxiliary supply for

the internal gate drive charge pump.

For typical MX16171D100/ MX16171S100 applications, the VS pin can be connected directly to the OUT pin. The capacitor value should be the lowest value that produces acceptable filtering of the voltage noise.

If VS is powered while IN is floating or grounded, then about 0.5 mA will leak from the VS pin into the IC and about 2mA will leak from the OUT pin into the IC.

OFF Pin

The OFF pin is a logic level input pin that is used to control the gate drive to the external MOSFET. The maximum operating voltage on this pin is 5.5V.

When the OFF pin is high, the MOSFET is turned off (independent of the sensed IN and OUT voltages). In this mode, load current will flow through the body diode of the MOSFET. The voltage difference between the IN pin and OUT pins will be approximately 700mV if the MOSFET is operating normally through the body diode.

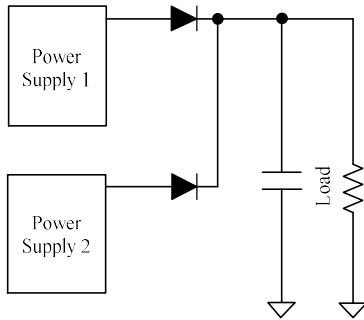
The OFF pin has an internal pulldown of 5 μ A (typical). If the OFF function is not required, the pin may be left open or connected to ground.

Application and Implementation

Application Information

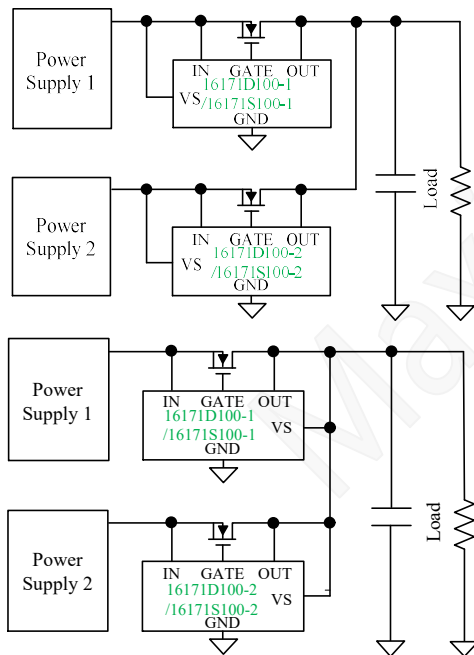
Systems that require high availability often use multiple, parallel-connected redundant power supplies to improve reliability. Schottky OR-ing diodes are typically used to connect these redundant power supplies to a common point at the load. The disadvantage of using OR-ing diodes is the forward voltage drop, which reduces the available voltage and the associated power losses as load currents increase. Using an N-channel MOSFET to replace the OR-ing diode requires a small increase in the level of complexity, but reduces, or eliminates, the need for diode heat sinks or large thermal copper area in circuit board layouts for high power applications.

100V Input High-Side OR-ing FET Controller



OR-ing with Diodes

The MX16171D100/ MX16171S100 is a positive voltage (that is, high-side) OR-ing controller that will drive an external N-channel MOSFET to replace an OR-ing diode. The voltage across the MOSFET source and drain pins is monitored by the MX16171D100/ MX16171S100 at the IN and OUT pins, while the GATE pin drives the MOSFET to control its operation based on the monitored source-drain voltage. The resulting behavior is that of an ideal rectifier with source and drain pins of the MOSFET acting as the anode and cathode pins of a diode respectively.



OR-ing With MOSFETs

MOSFET Selection

The important MOSFET electrical parameters are the maximum continuous Drain current I_D , the maximum Source current (that is, body diode) I_S , the maximum drain-to-source voltage $V_{DS(MAX)}$, the gate-to-source threshold voltage $V_{GS(TH)}$, the drain-to-source reverse breakdown voltage $V_{(BR)DSS}$, and the drain-to-source on resistance $R_{DS(ON)}$.

The maximum continuous drain current, I_D , rating must exceed the maximum continuous load current. The rating for the maximum current through the body diode, I_S , is typically rated the same as, or slightly higher than the drain current, but body diode current only flows while the MOSFET gate is being charged to $V_{GS(TH)}$.

$$\text{Gate Charge Time} = Q_g / I_{GATE(ON)}$$

1. The maximum drain-to-source voltage, $V_{DS(MAX)}$, must be high enough to withstand the highest differential voltage seen in the application. This would include any anticipated fault conditions.

2. The drain-to-source reverse breakdown voltage, $V_{(BR)DSS}$, may provide some transient protection to the OUT pin in low voltage applications by allowing conduction back to the IN pin during positive transients at the OUT pin.

3. The gate-to-source threshold voltage, $V_{GS(TH)}$, should be compatible with the MX16171D100/ MX16171S100 gate drive capabilities. Logic level MOSFETs, with $R_{DS(ON)}$ rated at $V_{GS(TH)}$ at 5V, are recommended, but sub-Logic level MOSFETs having $R_{DS(ON)}$ rated at $V_{GS(TH)}$ at 2.5V, can also be used.

4. The dominate MOSFET loss for the MX16171D100/ MX16171S100 active OR-ing controller is conduction loss due to source-to- drain current to the output load, and the $R_{DS(ON)}$ of the MOSFET. This conduction loss could be reduced by using a MOSFET with the lowest possible $R_{DS(ON)}$. However, contrary to popular belief, arbitrarily selecting a MOSFET based solely on having low $R_{DS(ON)}$ may not always give desirable results for several reasons:

1. Reverse transition detection. Higher $R_{DS(ON)}$ will provide increased voltage information to the MX16171D100/ MX16171S100 Reverse Comparator at a lower reverse current level. This will give an earlier MOSFET turnoff condition should the input voltage become shorted to ground. This will minimize any disturbance of the redundant bus.

2. Reverse current leakage. In cases where multiple input supplies are closely matched it may be possible for some small current to flow continuously through the MOSFET drain to source (that is, reverse) without activating the MX16171D100/ MX16171S100 Reverse Comparator. Higher $R_{DS(ON)}$ will reduce this reverse current level.

3. Cost. Generally, as the $R_{DS(ON)}$ rating goes lower, the cost of the MOSFET goes higher.

5. The dominate MOSFET loss for the MX16171D100/ MX16171S100 active OR-ing controller is conduction loss due to source-to- drain current to the output load, and the $R_{DS(ON)}$ of the MOSFET. This conduction loss could be reduced by using a MOSFET with the lowest possible $R_{DS(ON)}$. However, contrary to popular belief, arbitrarily selecting a MOSFET based solely on having low $R_{DS(ON)}$ may not always give desirable results for several reasons:

a. Selecting a MOSFET with an $R_{DS(ON)}$ that is too large will result in excessive power dissipation. Additionally, the MOSFET gate will be charged to the full value that the MX16171D100/ MX16171S100 can provide as it attempts to drive the Drain to Source voltage down to the $V_{SD(REG)}$ of 30mV typical. This increased Gate charge will require some finite amount of additional discharge time when the MOSFET needs to be turned off.

b. As a guideline, it is suggested that $R_{DS(ON)}$ be selected to provide at least 30mV, and no more than 100mV, at the nominal load current.

c. $(30mV / I_D) \leq R_{DS(ON)} \leq (100mV / I_D)$

d. The thermal resistance of the MOSFET package should also be considered against the anticipated dissipation in the MOSFET to ensure that the junction temperature (T_J) is reasonably well controlled, because the $R_{DS(ON)}$ of the MOSFET increases as the junction temperature increases.

6. $P_{DISS} = I_D^2 \times (R_{DS(ON)})$

7. Operating with a maximum ambient temperature ($T_{A(MAX)}$) of 35°C, a load current of 10 A, and an $R_{DS(ON)}$ of 10 mΩ, and desiring to keep the junction temperature under 100°C, the maximum junction-to-ambient thermal resistance rating (θ_{JA}) must be:

a. $R_{\theta JA} \leq (T_{J(MAX)} - T_{A(MAX)}) / (I_D^2 \times R_{DS(ON)})$

b. $R_{\theta JA} \leq (100^\circ C - 35^\circ C) / (10A \times 10A \times 0.01\Omega)$

c. $R_{\theta JA} \leq 65^\circ C/W$

Short Circuit Failure of an Input Supply

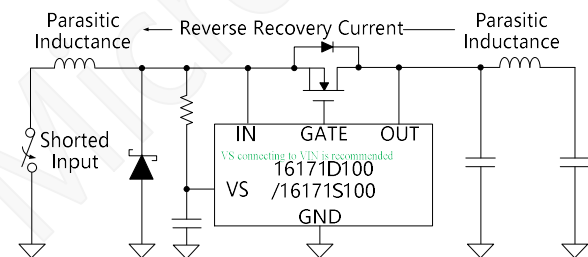
An abrupt 0Ω short circuit across the input supply will cause the highest possible reverse current to flow while the internal MX16171D100/ MX16171S100 control circuitry discharges the gate of the MOSFET. During this time, the reverse current is limited only by the $R_{DS(ON)}$ of the MOSFET, along with parasitic wiring resistances and inductances. Worst case instantaneous reverse current would be limited to:

$$I_{D(REV)} = (V_{OUT} - V_{IN}) / R_{DS(ON)} \quad (1)$$

The internal Reverse Comparator will react, and will start the process of discharging the Gate, when the reverse current reaches:

$$I_{D(REV)} = V_{SD(REV)} / R_{DS(ON)} \quad (2)$$

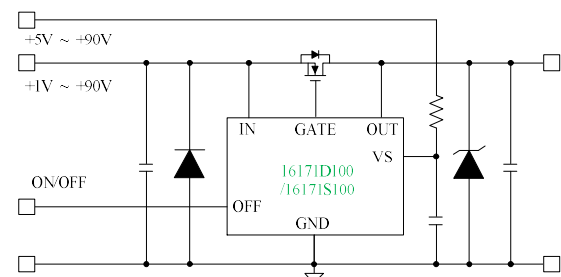
When the MOSFET is finally switched off, the energy stored in the parasitic wiring inductances will be transferred to the rest of the circuit. As a result, the MX16171D100/ MX16171S100 IN pin will see a negative voltage spike while the OUT pin will see a positive voltage spike. The IN pin can be protected by diode clamping the pin to GND in the negative direction. The OUT pin can be protected with a TVS protection diode, a local bypass capacitor, or both. In low voltage applications, the MOSFET drain to- source breakdown voltage rating may be adequate to protect the OUT pin (that is, $V_{IN} + V_{(BR)DSS(MAX)} < 40V$), but most MOSFET data sheets do not ensure the maximum breakdown rating, so this method should be used with caution.



Reverse Recovery Current Generates Spikes at V_{IN} and V_{OUT}

A Separate VS Supply for Low Vin Operation

In some applications, it is desired to operate MX16171D100/ MX16171S100 from low supply voltage. The MX16171D100/ MX16171S100 can operate with a 1V rail voltage, provides its VS pin is biased from 5V to 40. The detail of such application is depicted in the next figure.

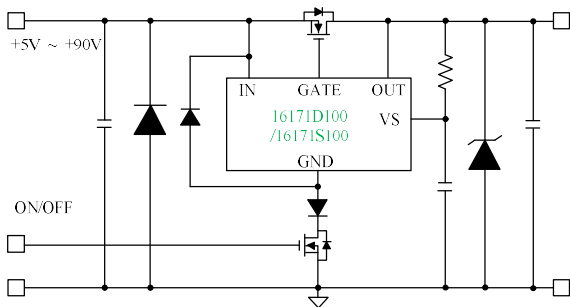


Reverse Input Voltage Protection with IQ Reduction

If VS is powered while IN is floating or grounded, then about 0.5mA will leak from the VS pin into the IC and about 3mA

will leak from the OUT pin into the IC. From this leakage, about 0.05mA will flow out of the IN pin and the rest will flow to ground. This does not affect long term reliability of the IC but may influence circuit design.

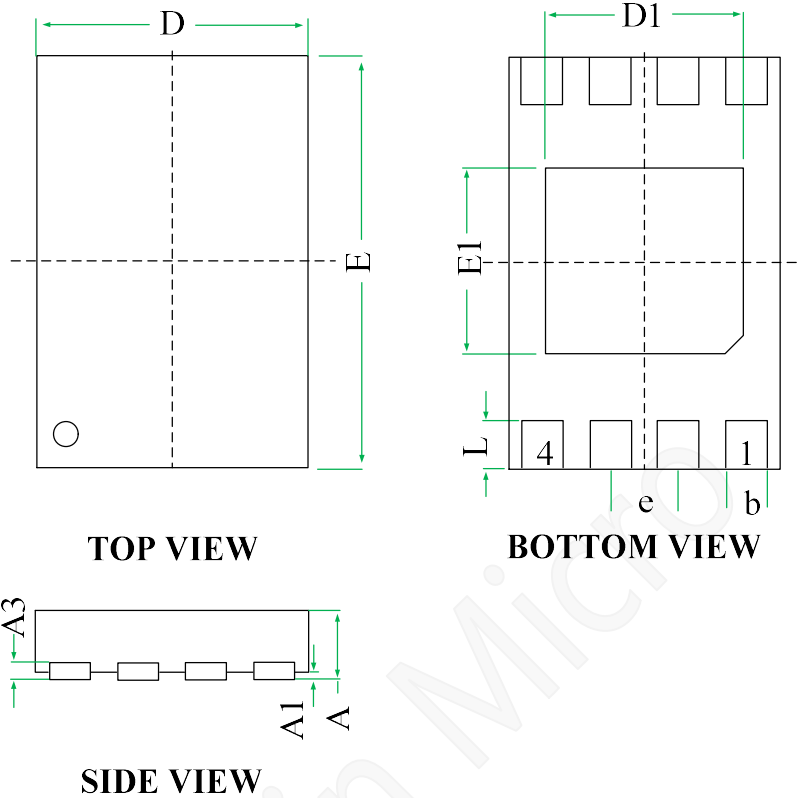
In battery powered applications, whenever MX16171D100/ MX16171S100 functionality is not needed, the supply to the MX16171D100/ MX16171S100 can be disconnected by turning OFF Q2, as shown in the following figure. This disconnects to the ground path of the MX16171D100/ MX16171S100 and eliminates the current leakage from the battery.



Reverse input voltage protection with IQ reduction schematic

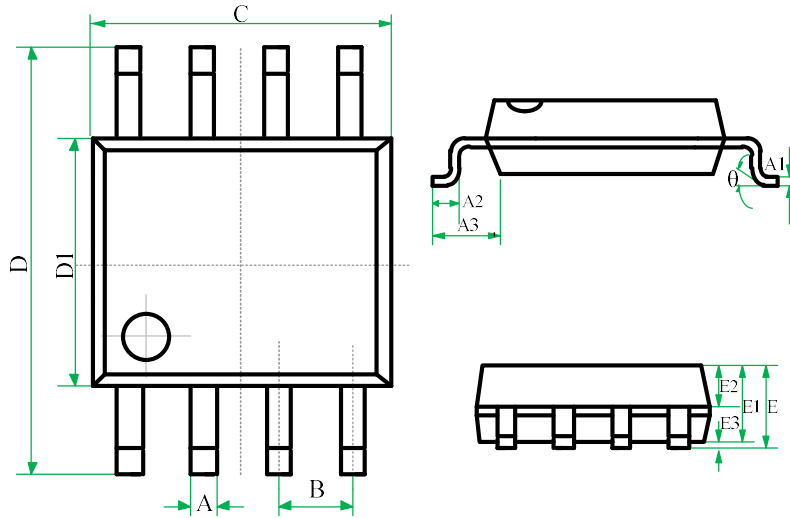
Package information

DFN2*3-8L



SYMBOL	MILLIMETERS		
	MIN	NOM	MAX
A	0.70	0.75	0.80
A1	0	0.02	0.05
A3	0.18	0.20	0.25
D	1.90	2.00	2.10
E	2.90	3.00	3.10
D1	1.40	1.50	1.60
E1	1.50	1.60	1.70
b	0.20	0.25	0.30
e		0.500BSC	
L	0.35	0.40	0.45

DFN2*3-8L for MX16171D100

SOP8


SYMBOL	MILLIMETERS			INCHES		
	MIN	NOM	MAX	MIN	NOM	MAX
A	0.39	-	0.48	0.0154	-	0.0189
A1	0.21	-	0.28	0.008	-	0.011
A2	0.50	-	0.80	0.020	-	0.031
A3	1.05BSC			0.041BSC		
B	1.27BSC			0.050BSC		
C	4.70	4.90	5.10	0.185	0.193	0.201
D	5.80	6.00	6.20	0.228	0.236	0.244
D1	3.70	3.90	4.10	0.146	0.154	0.161
E	-	-	1.75	-	-	0.069
E1	1.30	1.40	1.50	0.051	0.055	0.059
E2	0.60	0.65	0.70	0.024	0.026	0.028
E3	0.10	-	0.225	0.004	-	0.009
θ	0	-	8°	0	-	8°

SOP8 for MX16171S100

Restrictions on Product Use

- ◆ MAXIN micro is continually working to improve the quality and reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing MAXIN products, to comply with the standards of safety in making a safe design for the entire system, and to avoid situations in which a malfunction or failure of such MAXIN products could cause loss of human life, bodily injury or damage to property.
- ◆ In developing your designs, please ensure that MAXIN products are used within specified operating ranges as set forth in the most recent MAXIN products specifications.
- ◆ The information contained herein is subject to change without notice.

Version update record:

V10 The original version.

V11 new application with vs pin connect to VIN, added SOP8 package.

V12 new production MX16171S100 is added with SOP8L package.

V13 Update the test conditions for I_{VS} in the datasheet.

V14 updated the package information for MX16171D100

Maxin Micro